

# Modeling and Control of Distributed Parameter Systems: The Port Hamiltonian Approach

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# Outline

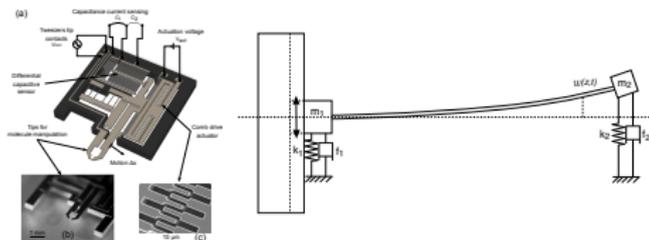
1. Context
2. Boundary controlled port Hamiltonian systems
3. Stabilization of BC PHS
4. Energy shaping for BC PHS
5. Application example



# Context : control of flexible structures

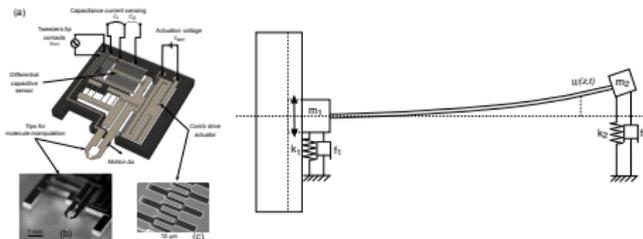


## ► Boundary controlled systems

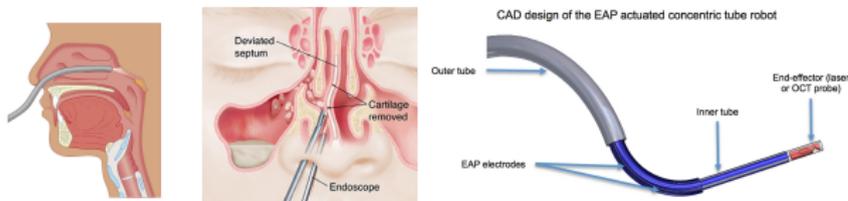


# Context : control of flexible structures

## ► Boundary controlled systems



## ► In-domain control of distributed parameter systems



- Exploration, imaging, diagnosis.
- Mini invasive surgery.
- Toward miniaturized and *smart* endoscopes.

# Context : port Hamiltonian systems



- ▶ Port Hamiltonian systems:
  - ▶ The state variables are chosen as the energy variables.
  - ▶ The links between the energy function and the system dynamics is made explicit through symmetries.
  - ▶ The boundary port variables are power conjugated.
- ▶ Energy shaping consists in using the physical properties of the system to derive efficient control laws with guaranteed performances (step further stabilization).
- ▶ "Easy" to extend to non linear or systems defined on higher dimensional spaces.



# Outline

Context

Boundary controlled port Hamiltonian systems

Stabilization of BC PHS

Energy shaping for BC PHS

Application example



# Port-Hamiltonian partial differential equations

Our model class are p.d.e.'s of the form

$$\frac{\partial x}{\partial t}(\zeta, t) = \left( P_1 \frac{\partial}{\partial \zeta} + P_0 \right) [\mathcal{H}x(\zeta, t)]$$

with

- ▶  $x(\zeta, t) \in \mathbb{R}^n$ ,  $\zeta \in [a, b]$ ,  $t \geq 0$
- ▶  $P_1$  is an invertible, symmetric real  $n \times n$ -matrix,
- ▶  $P_0$  is a skew-symmetric real  $n \times n$ -matrix,
- ▶  $\mathcal{H}(\zeta)$  is a symmetric, invertible  $n \times n$ -matrix with  $mI \leq \mathcal{H}(\zeta) \leq MI$  for some  $m, M > 0$ .



# Port-Hamiltonian partial differential equations



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The energy/Hamiltonian is defined as

$$H(t) = H(x(\cdot, t)) = \frac{1}{2} \int_a^b x(\zeta, t)^T \mathcal{H}(\zeta) x(\zeta, t) d\zeta.$$



# Boundary port variables



## Boundary port variables

Let  $\mathcal{H}x \in H^1(a, b; \mathbb{R}^n)$ . Then the boundary port variables are the vectors  $e_{\partial, \mathcal{H}x}, f_{\partial, \mathcal{H}x} \in \mathbb{R}^n$ ,

$$\begin{aligned} \begin{bmatrix} f_{\partial, \mathcal{H}x} \\ e_{\partial, \mathcal{H}x} \end{bmatrix} &= U \frac{1}{\sqrt{2}} \begin{bmatrix} P_1 & -P_1 \\ I & I \end{bmatrix} \begin{bmatrix} (\mathcal{H}x)(b) \\ (\mathcal{H}x)(a) \end{bmatrix} = R \begin{bmatrix} (\mathcal{H}x)(b) \\ (\mathcal{H}x)(a) \end{bmatrix} \\ &\Leftrightarrow \frac{1}{2} \frac{d}{dt} \|x\|_{\mathcal{H}}^2 = f_{\partial, \mathcal{H}x}^T e_{\partial, \mathcal{H}x}, \end{aligned}$$

Where

$$U^T \Sigma U = \Sigma, \quad \Sigma = \begin{bmatrix} 0 & I \\ I & 0 \end{bmatrix}, \quad \Sigma \in M_{2n}(\mathbb{R})$$





## Boundary controlled port Hamiltonian systems

Let  $W$  be a  $n \times 2n$  real matrix. If  $W$  has full rank and satisfies  $W\Sigma W^T \geq 0$ , then the system  $\frac{\partial x}{\partial t} = P_1 \frac{\partial}{\partial \zeta} (\mathcal{H}(\zeta)x)(t, \zeta) + (P_0 - G_0)\mathcal{H}(\zeta)x(t, \zeta)$  with input

$$u(t) = W \begin{bmatrix} f_{\partial, \mathcal{H}x}(t) \\ e_{\partial, \mathcal{H}x}(t) \end{bmatrix}$$

is a BCS on  $X$ . The operator  $\mathcal{A}x = P_1(\partial/\partial\zeta)(\mathcal{H}x) + (P_0 - G_0)\mathcal{H}x$  with domain

$$D(\mathcal{A}) = \left\{ \mathcal{H}x \in H^1(a, b; \mathbb{R}^n) \mid \begin{bmatrix} f_{\partial, \mathcal{H}x}(t) \\ e_{\partial, \mathcal{H}x}(t) \end{bmatrix} \in \ker W \right\}$$

generates a contraction semigroup on  $X$ .



# Input and output

Let  $\tilde{W}$  be a full rank matrix of size  $n \times 2n$  with  $\begin{bmatrix} W \\ \tilde{W} \end{bmatrix}$  invertible and let  $P_{W, \tilde{W}}$  be given by

$$P_{W, \tilde{W}} = \left( \begin{bmatrix} W \\ \tilde{W} \end{bmatrix} \Sigma \begin{bmatrix} W \\ \tilde{W} \end{bmatrix}^\top \right)^{-1} = \begin{bmatrix} W \Sigma W^\top & W \Sigma \tilde{W}^\top \\ \tilde{W} \Sigma W^\top & \tilde{W} \Sigma \tilde{W}^\top \end{bmatrix}^{-1}.$$

Define the output of the system as the linear mapping  $\mathcal{C} : \mathcal{H}^{-1}H^1(a, b; \mathbb{R}^n) \rightarrow \mathbb{R}^n$ ,

$$y = \mathcal{C}x(t) := \tilde{W} \begin{bmatrix} f_{\partial, \mathcal{H}x}(t) \\ e_{\partial, \mathcal{H}x}(t) \end{bmatrix}.$$

Then for  $u \in C^2(0, \infty; \mathbb{R}^k)$ ,  $\mathcal{H}x(0) \in H^1(a, b; \mathbb{R}^n)$ , and  $u(0) = W \begin{bmatrix} f_{\partial, \mathcal{H}x}(0) \\ e_{\partial, \mathcal{H}x}(0) \end{bmatrix}$  the following balance equation is satisfied:

$$\frac{1}{2} \frac{d}{dt} \|x(t)\|_{\mathcal{H}}^2 \leq \frac{1}{2} \begin{bmatrix} u(t) \\ y(t) \end{bmatrix}^\top P_{W, \tilde{W}} \begin{bmatrix} u(t) \\ y(t) \end{bmatrix}$$

# Input and output

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$$y = \mathcal{C}x(t) := \tilde{W} \begin{bmatrix} f_{\partial, \mathcal{H}x}(t) \\ e_{\partial, \mathcal{H}x}(t) \end{bmatrix}.$$

We choose  $W$  and  $\tilde{W}$  such that  $\begin{pmatrix} W \\ \tilde{W} \end{pmatrix} \Sigma \begin{pmatrix} W^T & \tilde{W}^T \end{pmatrix} = \Sigma$ .

In this particular case:

$$\frac{1}{2} \frac{d}{dt} \|x(t)\|_{\mathcal{H}}^2 \leq y^T(t)u(t). \quad (1)$$

# Static feedback control

## Impedance passive case

If the matrices  $W$  and  $\tilde{W}$  are selected such that  $P_{W, \tilde{W}} = \begin{bmatrix} 0 & I \\ I & 0 \end{bmatrix} = \Sigma$ , then the BCS fulfills

$$\frac{1}{2} \frac{d}{dt} \|x(t)\|_{\mathcal{L}}^2 \leq u^\top(t) y(t).$$

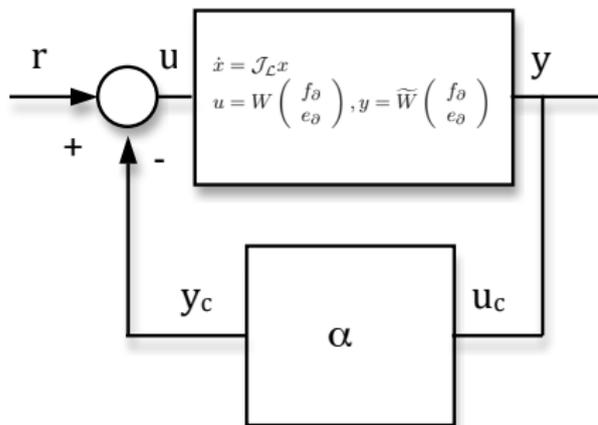


# Static feedback control

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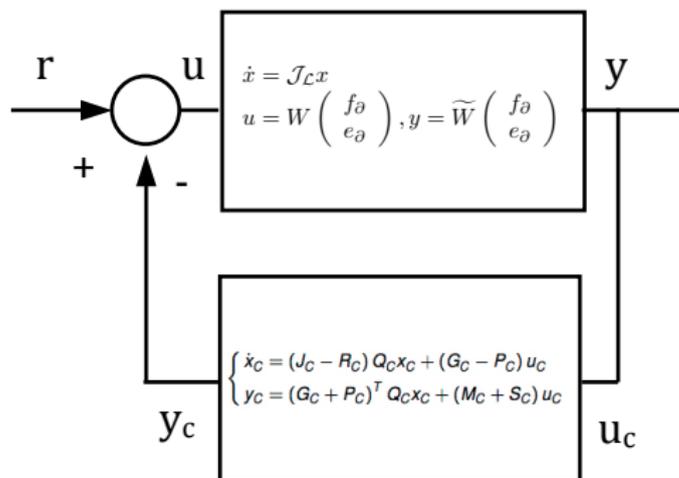
## Static controller

- ▶ Asymptotic stability:  
 $\alpha > 0$  (compactness condition)
- ▶ Exponential stability:  $\alpha$  st

$$(dE/dt) \leq -k \|(\mathcal{L}x)(t, b)\|_{\mathbb{R}}^2$$

where  $k > 0$ .

# Dynamic control



- ▶ Can we use passivity properties to design dynamic controllers ?
- ▶ What about closed loop trajectories ?
- ▶ Can we extend the energy shaping ideas to boundary controlled port Hamiltonian systems ?

# Energy shaping

We consider a dynamic controller of the form

$$\begin{cases} \dot{x}_C = (J_C - R_C) Q_C x_C + (G_C - P_C) u_C \\ y_C = (G_C + P_C)^T Q_C x_C + (M_C + S_C) u_C \end{cases} \quad (2)$$

where  $x_C \in \mathbb{R}^{n_C}$  and  $u_C, y_C \in \mathbb{R}^n$ , while  $J_C = -J_C^T$ ,  $M_C = -M_C^T$ ,  $R_C = R_C^T$ , and  $S_C = S_C^T$ , with this further condition satisfied:

$$\begin{pmatrix} R_C & P_C \\ P_C^T & S_C \end{pmatrix} \geq 0. \quad (3)$$

Interconnected to the boundary of the system

$$\begin{pmatrix} u \\ y \end{pmatrix} = \begin{pmatrix} 0 & -I \\ I & 0 \end{pmatrix} \begin{pmatrix} u_C \\ y_C \end{pmatrix} + \begin{pmatrix} u' \\ 0 \end{pmatrix}, \quad (4)$$

where  $u' \in \mathbb{R}^n$  is an additional control input.

# Existence of solutions

## Theorem

Let the open-loop BCS satisfy  $\frac{1}{2} \frac{d}{dt} \|x(t)\|_{\mathcal{L}}^2 = u(t)y(t)$  and consider the previous **passive** finite dimensional port Hamiltonian system. Then the power preserving feedback interconnection

$$u = r - y_c, y = u_c$$

with  $r \in \mathbb{R}^n$  the new input of the system is a BCS on the extended state space  $\tilde{x} \in \tilde{X} = X \times V$  with inner product  $\langle \tilde{x}_1, \tilde{x}_2 \rangle_{\tilde{X}} = \langle x_1, x_2 \rangle_{\mathcal{L}} + \langle v_1, Q_c v_2 \rangle_V$ . Furthermore, the operator  $\mathcal{A}_e$  defined by

$$\mathcal{A}_e \tilde{x} = \begin{bmatrix} \mathcal{J}\mathcal{L} & 0 \\ B_c \mathcal{C} & A_c \end{bmatrix} \begin{bmatrix} x \\ v \end{bmatrix},$$

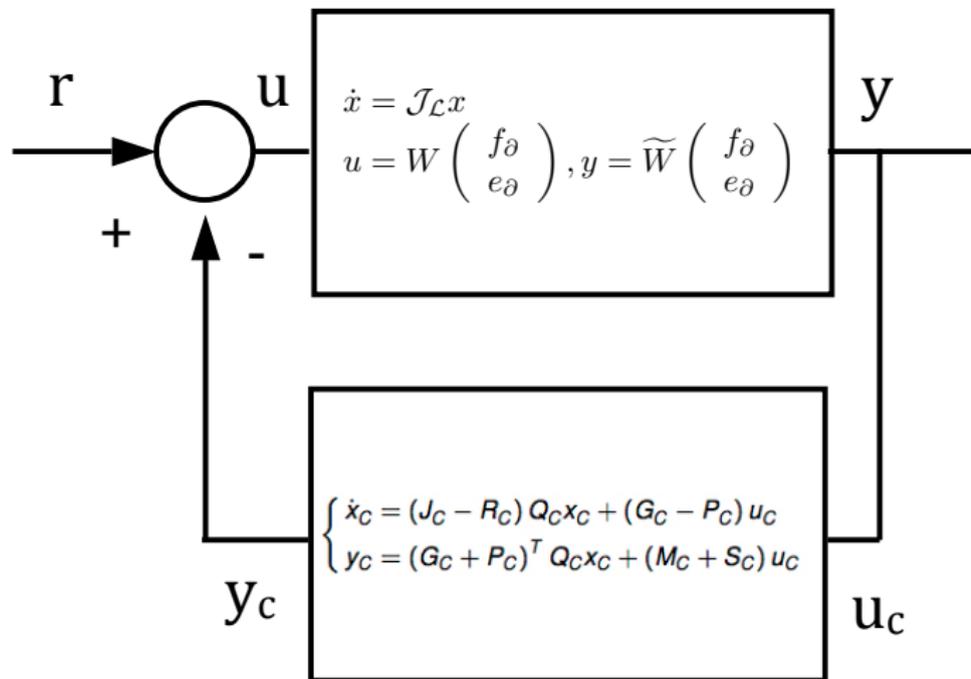
$$D(\mathcal{A}_e) = \left\{ \begin{bmatrix} x \\ v \end{bmatrix} \in \begin{bmatrix} X \\ V \end{bmatrix} \mid \mathcal{L}x \in H^N(a, b; \mathbb{R}^n), \begin{bmatrix} f_{\partial, \mathcal{L}x} \\ e_{\partial, \mathcal{L}x} \\ v \end{bmatrix} \in \ker \tilde{W}_D \right\}$$

where

$$\tilde{W}_D = [(W + D_c \tilde{W} \quad C_c)]$$

**generates a contraction semigroup** on  $\tilde{X}$ .

# Energy shaping



## Energy shaping

- ▶ Use of the structural invariants  $C$  (such that  $\dot{C} = 0$ ) of the form

$$C(x(t), x_c(t)) = \Gamma^T x_c(t) + \int_a^b \psi^T(\zeta)x(t, \zeta)dz = \kappa$$

to link the controller states to the system states.

- ▶ Choice of the controller energy function to "shape" the closed loop energy function as
$$H_{cl}(x(t), x_c(t)) = H(x(t)) + H_c(x_c(t)) = H(x(t)) + H_c(F(x(t)))$$
- ▶ Well known and very efficient for finite dimensional non linear systems.
- ▶ What about the linear boundary controlled infinite dimensional case ?



# Structural invariants

## Casimir functions

Consider the closed loop boundary control system with  $u' = 0$  then,

$$C(x(t), x_c(t)) = \Gamma^T x_c(t) + \int_a^b \psi^T(\zeta) x(t, \zeta) dz$$

is a Casimir function for this system **if and only if**  $\psi \in H^1(a, b; \mathbb{R}^n)$ ,

$$P_1 \frac{d\psi}{dz}(\zeta) + (P_0 + G_0)\psi(\zeta) = 0 \quad (5)$$

$$(J_C + R_C)\Gamma + (G_C + P_C)\tilde{W}R \begin{pmatrix} \psi(b) \\ \psi(a) \end{pmatrix} = 0 \quad (6)$$

$$(G_C - P_C)^T \Gamma + [W + (M_C - S_C)\tilde{W}]R \begin{pmatrix} \psi(b) \\ \psi(a) \end{pmatrix} = 0 \quad (7)$$



## Sketch of the proof

$C(x_e(t))$  is a Casimir function if and only if  $\frac{dC}{dt} = 0$  independently to the energy function,

$$\frac{dC}{dt} = \left\langle \frac{\delta C}{\delta x_e}, \frac{dx_e}{dt} \right\rangle_{L^2} \quad (8)$$

$$= \left\langle \frac{\delta C}{\delta x_e}, \mathcal{A}_e \mathcal{H}_e x_e \right\rangle_{L^2} \quad (9)$$

$$= \left\langle \mathcal{A}_e^* \frac{\delta C}{\delta x_e}, \mathcal{H}_e x_e \right\rangle_{L^2} + BC \quad (10)$$

$$(11)$$



## Proposition

Under the hypothesis that the Casimir functions exist, the closed-loop dynamics (when  $u = y_c + u'$ ) is given by :

$$\begin{aligned} \frac{\partial x}{\partial t}(t, \zeta) &= P_1 \frac{\partial}{\partial \zeta} \frac{\delta H_{cl}}{\delta x}(x(t))(\zeta) + (P_0 - G_0) \frac{\delta H_{cl}}{\delta x}(x(t))(\zeta) \\ u' &= W' R \begin{pmatrix} \left( \frac{\delta H_{cl}}{\delta x}(x) \right) (b) \\ \left( \frac{\delta H_{cl}}{\delta x}(x) \right) (a) \end{pmatrix} \end{aligned} \quad (12)$$

in which  $\delta$  denotes the variational derivative, while

$$\begin{aligned} H_{cl}(x(t)) &= \frac{1}{2} \|x(t)\|_{cl}^2 + \frac{1}{2} \left( \int_a^b \hat{\Psi}^T(\zeta) x(t, \zeta) dz \right)^T \times \\ &\quad \times \hat{\Gamma}^{-1} Q_C \hat{\Gamma}^{-T} \int_a^b \hat{\Psi}(\zeta)^T x(t, \zeta) dz \end{aligned} \quad (13)$$

and  $W'$  is a  $n \times 2n$  full rank, real matrix s.t.  $W' \Sigma W'^T \geq 0$ .

# Extension to systems with dissipation

## Proposition

The feedback law  $u = \beta(x) + u'$ , with  $u'$  an auxiliary boundary input, maps the original system into the target dynamical system

$$\begin{aligned} \frac{\partial x}{\partial t}(t, \zeta) &= P_1 \frac{\partial}{\partial \zeta} \frac{\delta H_d}{\delta x}(x(t))(\zeta) + (P_0 - G_0) \frac{\delta H_d}{\delta x}(x(t))(\zeta) \\ u'(t) &= WR \begin{pmatrix} \left( \frac{\delta H_d}{\delta x}(x(t)) \right) (b) \\ \left( \frac{\delta H_d}{\delta x}(x(t)) \right) (a) \end{pmatrix} \end{aligned} \quad (14)$$

with  $H_d(x) = H(x) + H_a(x)$ , provided that

$$P_1 \frac{\partial}{\partial \zeta} \frac{\delta H_a}{\delta x}(x) + (P_0 - G_0) \frac{\delta H_a}{\delta x}(x) = 0 \quad (15)$$

$$\beta(x) + WR \begin{pmatrix} \left( \frac{\delta H_a}{\delta x}(x) \right) (b) \\ \left( \frac{\delta H_a}{\delta x}(x) \right) (a) \end{pmatrix} = 0. \quad (16)$$

# Energy shaping

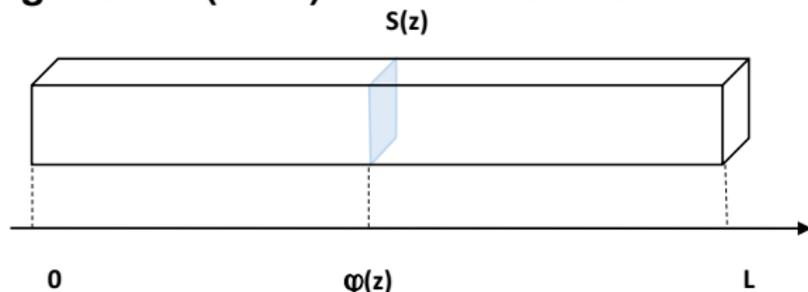


With the dynamic extension or state feedback we have been able to shape a part of the closed loop energy function. It remains to prove that the closed loop system is asymptotically stable.

- ▶ We have to consider additional damping injection.
- ▶ Exponential stabilisation is not possible as "exponential stability of the controller + direct feedforward term" are necessary → no Casimir function.



## Example: longitudinal (axial) vibration of a beam



State variables : deformation and linear momentum density

$$\varepsilon(t, \zeta) = \frac{\partial \varphi}{\partial \zeta}(t, \zeta), \quad p(t, \zeta) = \rho S(\zeta)v(t, \zeta) \quad (17)$$

Material's deformation is considered linear (Hooke's law) :

$$\rho S(\zeta) \frac{\partial^2 \varphi}{\partial t^2}(t, \zeta) = \frac{\partial}{\partial \zeta} \left[ ES(\zeta) \frac{\partial \varphi}{\partial \zeta}(t, \zeta) \right] - D \frac{\partial \varphi}{\partial t}(t, \zeta) d\zeta$$

The energy is given by (kinetic+potential):

$$H(p(t, \zeta), \varepsilon(t, \zeta)) = \frac{1}{2} \int_0^L \left[ \frac{p^2(t, \zeta)}{\rho S(\zeta)} + ES(\zeta)\varepsilon^2(t, \zeta) \right] d\zeta$$

## Example: longitudinal (axial) vibration of a beam

From:

$$H(p(t, \zeta), \varepsilon(t, \zeta)) = \frac{1}{2} \int_0^L \left[ \frac{p^2(t, \zeta)}{\rho S(\zeta)} + ES(\zeta)\varepsilon^2(t, \zeta) \right] d\zeta$$

We define the co-energy variables:

$$\begin{aligned}\sigma_S(t, \zeta) &= \frac{\delta H}{\delta \varepsilon}(\varepsilon(t, \zeta)) = ES(\zeta)\varepsilon(t, \zeta) = S(\zeta)\sigma(t, \zeta) \\ v(t, \zeta) &= \frac{\delta H}{\delta p}(p(t, \zeta)) = \frac{p(t, \zeta)}{\rho S(\zeta)} = \frac{\partial \varphi}{\partial t}(t, \zeta)\end{aligned}$$

Then:

$$\frac{\partial}{\partial t} \left( \rho S(\zeta) \frac{\partial \varphi}{\partial t}(t, \zeta) \right) = \frac{\partial}{\partial \zeta} \left[ ES(\zeta) \frac{\partial \varphi}{\partial \zeta}(t, \zeta) \right] - D \frac{\partial \varphi}{\partial t}(t, \zeta)$$

with

$$\frac{\partial}{\partial t} \left( \frac{\partial \varphi}{\partial \zeta}(t, \zeta) \right) = \frac{\partial}{\partial \zeta} \left( \frac{\partial \varphi}{\partial t}(t, \zeta) \right)$$

## Example: longitudinal (axial) vibration of a beam



The port-Hamiltonian formulation of the system is then

$$\frac{\partial}{\partial t} \begin{pmatrix} \varepsilon(t, \zeta) \\ p(t, \zeta) \end{pmatrix} = \begin{pmatrix} 0 & \frac{\partial}{\partial \zeta} \\ \frac{\partial}{\partial \zeta} & -D \end{pmatrix} \begin{pmatrix} ES(\zeta) & 0 \\ 0 & \frac{1}{\rho S(\zeta)} \end{pmatrix} \begin{pmatrix} \varepsilon(t, \zeta) \\ p(t, \zeta) \end{pmatrix}$$

which is in the form :

$$\frac{\partial x}{\partial t}(t, \zeta) = P_1 \frac{\partial}{\partial \zeta} (\mathcal{H}(\zeta)x(t, \zeta)) + (P_0 - G_0)\mathcal{H}(\zeta)x(t, \zeta) \quad (18)$$

with  $P_0 = 0$  and

$$P_1 = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \quad G_0 = \begin{pmatrix} 0 & 0 \\ 0 & D \end{pmatrix} \quad \mathcal{H}(\zeta) = \begin{pmatrix} ES(\zeta) & 0 \\ 0 & \frac{1}{\rho S(\zeta)} \end{pmatrix}$$



# Input and output

The boundary port variables are

$$\begin{pmatrix} f_{\partial} \\ e_{\partial} \end{pmatrix} = \frac{1}{\sqrt{2}} \begin{pmatrix} v(L) - v(0) \\ \sigma_S(L) - \sigma_S(0) \\ \sigma_S(L) + \sigma_S(0) \\ v(L) + v(0) \end{pmatrix}$$

The boundary input and output are selected as

$$u(t) = \begin{pmatrix} v(t, 0) \\ \sigma_S(t, L) \end{pmatrix} \quad y(t) = \begin{pmatrix} -\sigma_S(t, 0) \\ v(t, L) \end{pmatrix} \quad (19)$$

which can be derived choosing  $W$  and  $\tilde{W}$  such that:

$$W = \frac{1}{\sqrt{2}} \begin{pmatrix} -1 & 0 & 0 & 1 \\ 0 & 1 & 1 & 0 \end{pmatrix} \quad \tilde{W} = \frac{1}{\sqrt{2}} \begin{pmatrix} 0 & 1 & -1 & 0 \\ 1 & 0 & 0 & 1 \end{pmatrix}$$

The energy balance is then :

$$\frac{dH}{dt}(t) = - \int_0^L Dv^2(t, \zeta) d\zeta + y^T(t)u(t) \leq y^T(t)u(t).$$



## Lossless case : Approach based on structural invariants

We consider a dynamic controller with  $n_C = 2$ ,  $R_C = P_C = M_C = S_C = 0$ ,  $G_C = I$  and

$$J_C = \begin{pmatrix} 0 & I \\ -I & 0 \end{pmatrix},$$

which implies that the closed-loop system is characterized by the following Casimir functions:

$$C_1(\xi_1(t), \varepsilon(t, \cdot)) = \xi_1(t) - \int_0^L \varepsilon(t, \zeta) d\zeta$$
$$C_2(\xi_2(t), p(t, \cdot)) = \xi_2(t) - \int_0^L p(t, \zeta) d\zeta.$$

The controller Hamiltonian is chosen such that

$$\hat{H}_c(\xi_1, \xi_2) = \frac{1}{2} \Xi_1 \xi_1^2 + \frac{1}{2} \Xi_2 \xi_2^2 \quad (20)$$

# Approach based on structural invariants



The closed loop energy function is:

$$H_{cl}(\varepsilon, p) = \frac{1}{2} \int_0^L \left[ \frac{p^2}{\rho S(\zeta)} + ES(\zeta)\varepsilon^2 \right] d\zeta + \frac{1}{2} \Xi_1 \left( \int_0^L \varepsilon d\zeta \right)^2 + \frac{1}{2} \Xi_2 \left( \int_0^L p d\zeta \right)^2 \quad (21)$$

and the control is of the form

$$u = -y_c = -G_c \delta H_c = - \begin{pmatrix} \Xi_2 & 0 \\ 0 & \Xi_1 \end{pmatrix} \begin{pmatrix} \int_0^L p d\zeta \\ \int_0^L \varepsilon d\zeta \end{pmatrix}$$



# System with dissipation

Due to the dissipation  $D \neq 0$ , the energy-Casimir method cannot be applied. The closed loop energy function cannot be shaped in the  $p$  coordinate.

Admissible  $H_a$  :

$$\hat{H}_a(\xi_1, \xi_2) = \frac{1}{2}\Xi_1\xi_1^2 + \frac{1}{2}\Xi_2\xi_2^2$$

with

$$\xi_1(\varepsilon(t, \cdot)) = \int_0^L \varepsilon(t, \zeta) d\zeta \tag{22}$$

$$\xi_2(\varepsilon(t, \cdot), p(t, \cdot)) = \int_0^L [D(L-z)\varepsilon(t, \zeta) + p(t, \zeta)] d\zeta$$

Leading to

$$u = - \begin{pmatrix} \Xi_2 & 0 \\ 0 & \Xi_1 \end{pmatrix} \begin{pmatrix} \int_0^L [D(L-z)\varepsilon(t, \zeta) + p(t, \zeta)] d\zeta \\ \int_0^L \varepsilon d\zeta \end{pmatrix}$$



# Achievable performances

We consider now that  $D = 0$ , all parameters equal 1 (simulations are provided considering a finite volume approximation)

$$u(t) = \begin{pmatrix} v(t, 0) \\ \sigma_S(t, L) \end{pmatrix} = \begin{pmatrix} 0 \\ \bar{u}(t) \end{pmatrix} \quad y(t) = \begin{pmatrix} -\sigma_S(t, 0) \\ v(t, L) \end{pmatrix} = \begin{pmatrix} \tilde{y}(t) \\ \bar{y}(t) \end{pmatrix}$$

and we plot the position at the end point of the system.

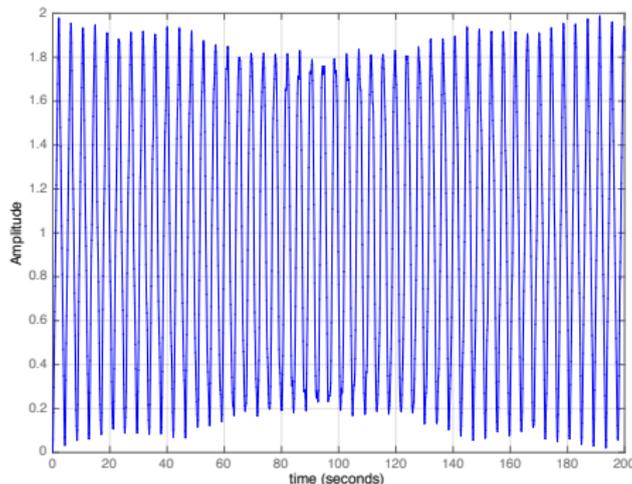


Figure: Open loop step response.

# Simulation

We first consider the static feedback case *i.e.* when pure dissipation is added at the boundary:

$$u_2 = -k_d y_2$$

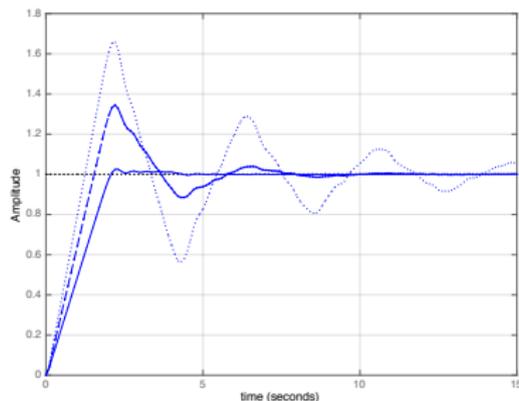


Figure: Step response of the closed loop system with pure dissipation term.



# Simulation

In a second instance we consider the control law devoted to energy shaping in addition to a pure dissipation term:

$$u = -k_c (x_{22} - x_{01}) - k_d \dot{x}_{22}$$

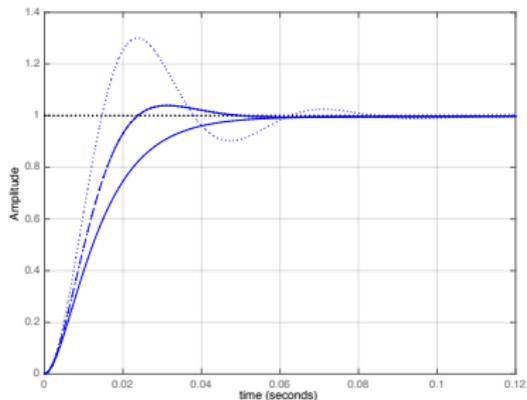


Figure: Step response of the closed loop system with state feedback.



# Conclusion and future work



- ▶ A large class of boundary control systems are asymptotically (exponentially) stable if they are interconnected in a power preserving manner with an (input strictly passive and) exponentially stable finite dimensional linear controller.
- ▶ Stability established for static control of BCS has been extended to the case of dynamic boundary control.
- ▶ These results can be used for control design.



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## Ongoing and future work

- ▶ Generalization to 2D and 3D systems.
- ▶ Extension to non-linear PDEs
- ▶ Constructive methods for control design.



Thank you for your attention !

